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# ROCKY MOUNTAIN FOREST AND RANGE EXPERIMENT STATION

## The Rocky Mountain Millivolt Integrator for Use with Solar Radiation Sensors

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Electronic integration of a radiometer's millivolt signal is a practical and accurate means of obtaining hourly, daily, weekly, or long-term radiation values. Our integrator consists of four printed circuit boards, a synchronous bi-directional stepper motor, and 5-decade counter. Each integrator is calibrated to match the millivolt output of the radiation sensor, so that the counter reads directly in langleys. The totalizing of a signal from a typical net radiometer with a 6.20mv/langley output) would be within  $\pm 1$  percent over most of the positive signal range, but could be 5 percent too low at night when the sensor output is negative.

Keywords: Solar radiation, instrumentation, electronic equipment.

Electronic integration is a practical and accurate means of obtaining radiation values over any time period greater than about a minute. Most often we are interested in daily, monthly, or yearly values. Reducing radiation data from strip charts is time consuming and inaccurate.

The Rocky Mountain millivolt integrator, designed for either a solar pyranometer (fig. 1) or net radiometer, is calibrated to match the millivolt output of the sensor so that it reads directly in langelys.

There are several integrators or totalizers described in the literature. A list of references on the subject is given at the end of this Note. Tanner (1965) did an excellent job of describing the various types of integrators available at that time. Since 1965, advances in integrated circuits have allowed an increase in instrumental accuracy and a decrease in power requirements. The

integrator described here is unique in that it employs a bi-directional stepping motor to drive a counter.



Figure 1.--Voltage integrator sums millivolt signal from pyranometer directly in langelys and displays it on a 5-decade counter. It can be read over any time period of interest from 5 minutes to several days.

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Integrators that employ the coulometer, Solion, or other types of electrolytics are probably the most economical, but they are also the least accurate, mainly because of readout difficulties. A major drawback to many of the integrators discussed in the literature is the large DC current drain, or the need for AC current.

The Rocky Mountain integrator combines accuracy, low cost, and low power requirements. The cost of the instrument we built was approximately \$150. Current drain averaged slightly over 300 milliamps.

## Construction and Principle of Operation

The Rocky Mountain millivolt integrator consists of four printed circuit boards (three when used with a pyranometer), a synchronous bi-directional stepper motor, and a 5-decade counter. A block diagram and an overall schematic of the integrator are given in figure 2. The four printed circuit boards are illustrated schematically in figures 3, 4, and 5, and photographically in figure 6.

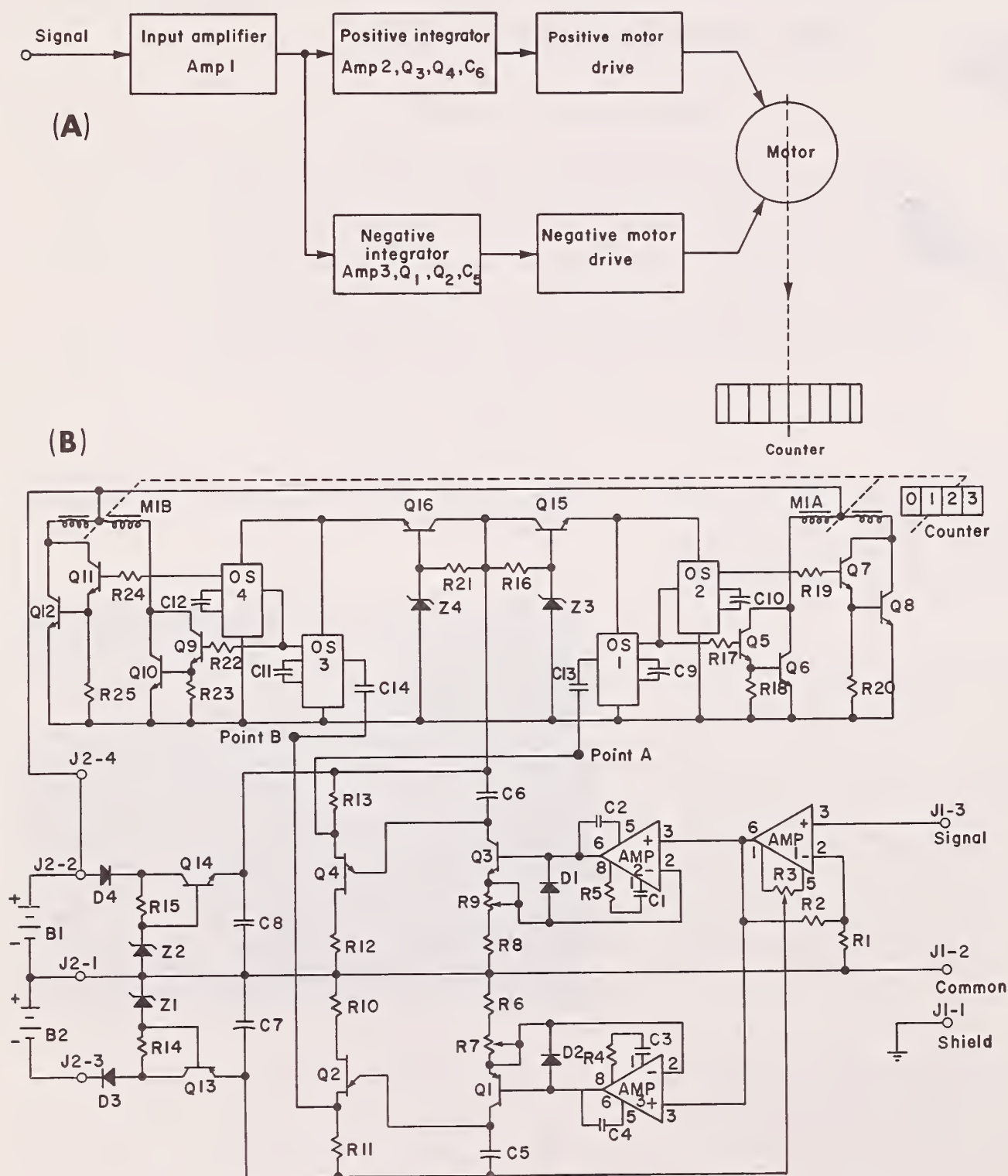


Figure 2.--  
Rocky Mountain  
millivolt  
integrator:

A, block  
diagram;

B, schematic  
diagram.



The circuit diagram shows a differential amplifier with two 2N4237 JFETs and two 2N4234 JFETs. The circuit is powered by +9V, 0V, and -9V rails. Each JFET is biased with a 47μF 15V capacitor. The output of each JFET is taken from the drain, which is connected to a 220Ω 1/2W resistor and a 757A diode. The diodes are connected to +12V and -12V rails.

The diagram shows a motor speed control circuit. It features two MC85IP comparators. The first comparator's non-inverting input (pin 4) is connected to a +5V supply through a 10k resistor. Its inverting input (pin 6) is connected to a 3.3μF capacitor, which is in turn connected to a 1.8k resistor leading to the base of a 2N4237 transistor. The output of this transistor (pin 10) is connected to the base of a 2N2222 transistor. The emitter of this 2N2222 is grounded, and its collector is connected to the base of another 2N4237 transistor. The second comparator's non-inverting input (pin 4) is connected to a +9V supply through a 10k resistor. Its inverting input (pin 6) is connected to a 3.3μF capacitor, which is in turn connected to a 1.8k resistor leading to the base of a 2N4237 transistor. The output of this transistor (pin 10) is connected to the base of a 2N2222 transistor. The emitter of this 2N2222 is grounded, and its collector is connected to the base of another 2N4237 transistor. The two 2N4237 transistors are connected in series to drive a motor, which is connected between their collectors and a +12V supply. The motor's other terminal is connected to ground. Various resistors (1.8k, 5.6k, 390Ω) and capacitors (3.3μF) are used throughout the circuit for timing and signal conditioning.

The diagram illustrates a motor speed control circuit. It features two 741 op-amps and two 709 comparators. The first 741 op-amp (top) has its non-inverting input (pin 3) connected to a 'Sig.' input and its inverting input (pin 2) connected to a feedback network consisting of a 20K resistor and a 1% resistor. Its output (pin 6) is connected to the non-inverting input (pin 3) of the first 709 comparator. The second 741 op-amp (bottom) has its non-inverting input (pin 3) connected to a '0v' input and its inverting input (pin 2) connected to a feedback network consisting of a 10K resistor and a 1% resistor. Its output (pin 6) is connected to the non-inverting input (pin 3) of the second 709 comparator. The first 709 comparator (top) has its inverting input (pin 2) connected to a feedback network consisting of a 47pF capacitor and a 1.2K resistor. Its output (pin 8) is connected to the base of a 2N3711 transistor. The second 709 comparator (bottom) has its inverting input (pin 2) connected to a feedback network consisting of a 47pF capacitor and a 1.2K resistor. Its output (pin 8) is connected to the base of a 2N4062 transistor. Both transistors are connected to a common emitter/ground point. The collector of the 2N3711 transistor is connected to a +9V supply through a 10μF capacitor and a 25V diode. The collector of the 2N4062 transistor is connected to a -9V supply through a 10μF capacitor and a 25V diode. The circuit also includes a 0V reference point and two output pulses: 'Motor up pulse' and 'Motor down pulse'. Various other components like resistors (2K, 10K, 20K, 240Ω, 200Ω, 2.7K, 10Ω), capacitors (0.1μF, 0.01μF, 47pF), and diodes (1N914, 50V) are used throughout the circuit.

Figure 6.-- The printed circuit board layout used in the Rocky Mountain integrator:

A, power supply;

B, amplifier;

C, motor driver;

D, the completed unit.



The integrator is similar to several other integrators defined as "relaxation oscillators" by Tanner (1965). It is unique in that it uses a bi-directional stepping motor to drive a counter. The stepping motor is driven by pulses from two integrating circuits, one for positive and one for negative input signals. The two integrating circuits are calibrated separately, thus allowing the output of a net radiometer to be totalized.

In figure 2B, the input amplifier, Amp 1, is a low-drift integrated-circuit operational amplifier with internal frequency compensation. It is connected in the non-inverting mode which presents a high input impedance to the input signal. The gain is set at 1000 by the ratio  $R2/R1$ .  $R3$  is a balance adjustment used to set the output of Amp 1, with zero signal, to compensate for offsets in Amp 1, Amp 2, and Amp 3.

The positive integrator is composed of Amp 2, Q3, and Q4. Amp 2 and Q3 charge C6 with a current equal to  $(1000 \times \text{Sig})/(R8 + R9)$ . When the voltage across C6 reaches the peak point of the complementary unijunction, Q4, the unijunction conducts and discharges C6. A pulse is fed to the shaping and driving circuits consisting of monostable multivibrators, OS1 and OS2, and associated transistors. The negative integrator consists of Amp 3, Q1, and Q2, which drive OS3 and OS4 and their transistors.

### Calibration

An accurate voltage divider was developed to insure proper calibration of the voltage inte-

grator for a given radiometer. This allows long-term calibration checks with constant ( $\pm 0.0025$  percent) millivolt inputs within the range of the output level from a radiation sensor.

The bench setup used in calibrating the integrator is shown in figure 7. The voltage divider is used to simulate the output from the radiometer that will be used with a particular integrator. A precision millivolt potentiometer is necessary to adjust the output from the divider to the desired levels that will make the integrator read directly in langley.

Because the integrator output is linear with respect to input, only one point of each polarity of input is necessary for calibration. This point was arbitrarily chosen to represent 1.0 langley.

When the input voltage is set at the level representing 1.0 langley, the time interval between pulses from the integrating circuits (Point A for positive input, Point B for negative, fig. 2B) is set with  $R9$  and  $R7$ , respectively. The time interval should be 100 milliseconds at 1.0 langley and 400 milliseconds when checked at 0.25 langley. Using a frequency counter with period averaging, the period can be set to 1/2 percent on the high level with 2 percent accuracy on the low level.

Long-term precision of the integrator was tested by supplying a constant input voltage from the divider over a period of 5 days. The error in voltage integration was only -0.9 percent on the negative input side and  $\pm 0.6$  percent on the positive input side.

Calibration tests under temperature conditions ranging from 75° to 108°F are presented in figure 8. Because wet-cell batteries are required

Figure 7.--To accurately calibrate the voltage integrator for a given radiometer, an accurate voltage divider was developed by the Rocky Mountain Station. This allows long-term calibration checks with constant ( $\pm 0.0025$  percent) millivolt inputs that simulate the output from a radiation sensor.



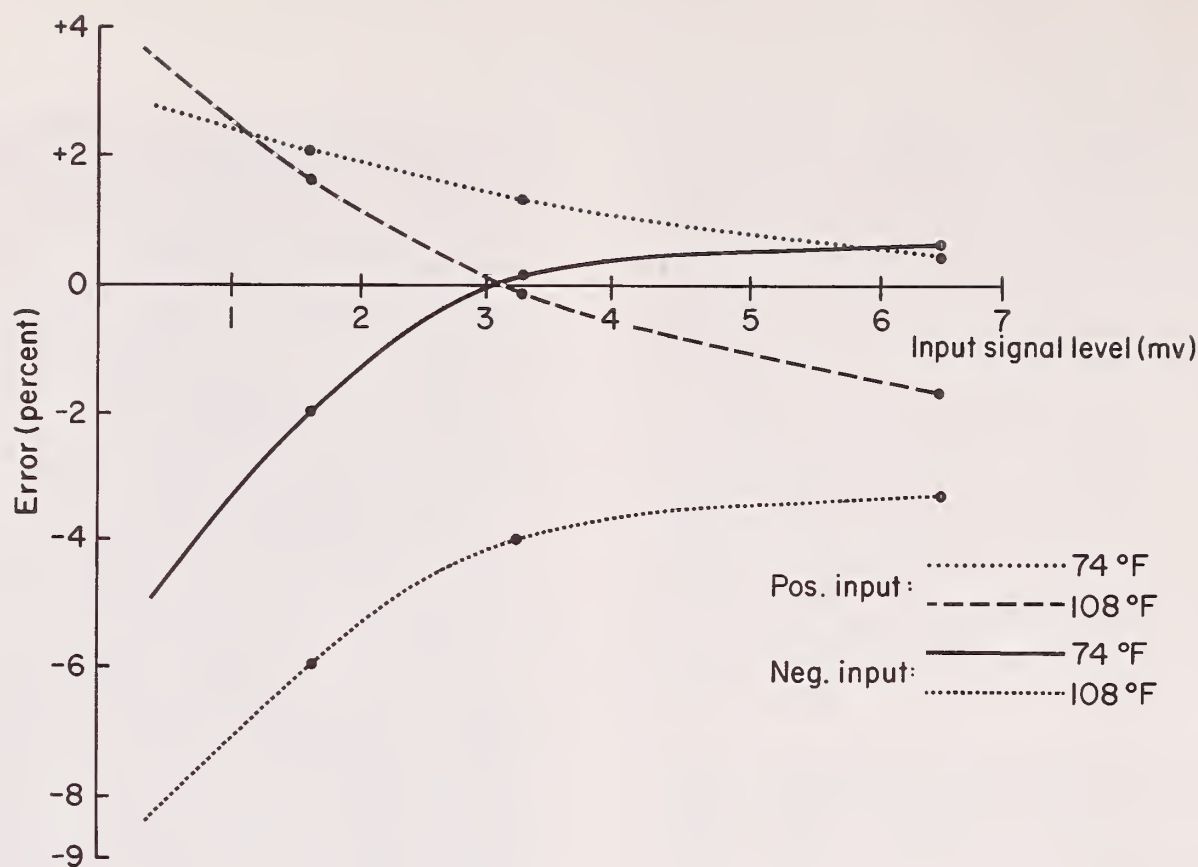


Figure 8.--Variations in the calibration of the RM millivolt integrator with temperature and signal level.

for a power source, a heated shelter is required when freezing temperatures are encountered. Maintaining a shelter environment of 75° to 100° F would minimize the integrator errors caused by temperature drift. Negative net radiation amounts to only 10 to 20 percent of the daily total, therefore a 5 to 6 percent nighttime integrator error is comparable to a 1 to 2 percent daytime error. If nighttime values are of interest by themselves, the shelter temperature becomes more critical in order to minimize the negative integrator error.

### References

- Brown, D. P., and R. A. Harvey.  
1961. Solar- and sky-radiation integrator. *Am. Meteorol. Soc. Bull.* 42: 325-332.
- Funk, J. P.  
1960. Sensitive and simple integrator. *J. Sci. Inst.* 37: 276-278.
- Goodell, B. C.  
1962. An inexpensive totalizer of solar and thermal radiation. *J. Geophys. Res.* 67: 1383-1387.
- Hanks, R. J., and H. R. Gardner.  
1964. Portable integrator for net radiation, total radiation, and soil heat flow. *Soil Sci. Soc. Am. Proc.* 28: 449-450.
- Monteith, J. L., and G. Szeicz.  
1960. The performance of a Gunn-Bellani radiation integrator. *Q. J. Roy. Meteorol. Soc.* 86: 91-94.
- Schoffer, P., and V. E. Suomi.  
1961. A direct current motor integrator for radiation measurements. *Solar Energy* 5: 29-32.
- Tanner, C. B.  
1965. Basic instrumentation and measurements for plant environment and micrometeorology. *Wis. Univ., Dep. Soil Sci., Soils Bull.* 6, var. p.
- Tanner, C. B., G. T. Thurtell, and J. B. Swan.  
1963. Integration systems using a commercial coulometer. *Soil Sci. Soc. Am. Proc.* 27: 478-481.
- Thompson, Owen E.  
1965. Low cost, portable millivolt integrator. *J. Appl. Meteorol.* 4(2): 289-291.
- Thurtell, G. W., and C. B. Tanner.  
1964. Electronic integrator for micrometeorological data. *J. Appl. Meteorol.* 3: 198-202.
- Turner, Duane H.  
1966. A highly stable electronic integrator for solar radiation measurements. *J. Appl. Meteorol.* 5: 895-896.



## Parts List<sup>2</sup>

B1 Battery, 12v lead-acid	Q1 Transistor 2N4062
B2 Battery, 12v lead-acid	Q2 Transistor 2N3711
C1 Capacitor 47pf	Q3 Transistor 2N3711
C2 Capacitor 0.1μf	Q4 Transistor D5K1 (G.E.)
C3 Capacitor 47pf	Q5 Transistor 2N2222A
C4 Capacitor 0.1μf	Q6 Transistor 2N4327
C5 Capacitor 10μf, 15v	Q7 Transistor 2N2222A
C6 Capacitor 10μf, 15v	Q8 Transistor 2N4237
C7 Capacitor 47μf, 15v	Q9 Transistor 2N2222A
C8 Capacitor 47μf, 15v	Q10 Transistor 2N4237
C9 Capacitor 3.3μf, 15v	Q11 Transistor 2N2222A
C10 Capacitor 3.3μf, 15v	Q12 Transistor 2N4237
C11 Capacitor 3.3μf, 15v	Q13 Transistor 2N4234
C12 Capacitor 3.3μf, 15v	Q14 Transistor 2N4237
C13 Capacitor 0.1μf	Q15 Transistor 2N4237
C14 Capacitor 0.1μf	Q16 Transistor 2N4237
D1 Diode 1N914	Unless otherwise stated, all resistors are $\pm 5\%$ , 1/4W
D2 Diode 1N914	
D3 Diode 1N4005	
D4 Diode 1N4005	
J1 Jack 91-855 (Amphenol) (Mates 91-854)	R1 Resistor 20Ω, $\pm 1\%$ , 1/4W
J2 Jack 91-859 (Amphenol) (Mates 91-858)	R2 Resistor 2KΩ, $\pm 1\%$ , 1/4W
M1 Stepping motor K-44135-P2; 50:1 gear reduction (A. W. Haydon)	R3 Variable resistor 10K, Pot.; Bourns #3280P-1-103
Counter A1141 25-005 (Veeder-Root)	R4 Resistor 1.2KΩ
Edge connectors 251-20A-30 (Cinch-Jones)	R5 Resistor 1.2KΩ
	R6 Resistor 2.7Ω (Selected for proper range of calibration adjustment)
	R7 Variable resistor 2.7Ω, Pot.; Bourns #3280P-1-202
	R8 Resistor 240Ω (Selected for proper range of calibration adjustment)
	R9 Variable resistor 2KΩ, Pot.; Bourns #3280P-1-202
	R10 Resistor 200Ω

<sup>2</sup> The use of trade and company names is for the benefit of the reader; such use does not constitute an official endorsement or approval of any service or product by the U. S. Department of Agriculture to the exclusion of others that may be suitable.

R11 Resistor 10 $\Omega$

R12 Resistor 200 $\Omega$

R13 Resistor 10 $\Omega$

R14 Resistor 220 $\Omega$ ,  $\pm$  5%, 1/2W

R15 Resistor 220 $\Omega$ ,  $\pm$  5%, 1/2W

R16 Resistor 390 $\Omega$ ,  $\pm$  5%, 1/2W

R17 Resistor 1.8K $\Omega$

R18 Resistor 5.6K $\Omega$

R19 Resistor 1.8K $\Omega$

R20 Resistor 5.6K $\Omega$

R21 Resistor 390 $\Omega$ ,  $\pm$  5%, 1/2W

R22 Resistor 1.8K $\Omega$

R23 Resistor 5.6K $\Omega$

R24 Resistor 1.8K $\Omega$

R25 Resistor 5.6K $\Omega$

Z1 Zener diode 1N4739A

Z2 Zener diode 1N4739A

Z3 Zener diode 1N4734A

Z4 Zener diode 1N4734A

Amp 1 Operational amplifier U5B7741312 (Fairchild)

Amp 2 Operational amplifier U5B770931X (Fairchild)

Amp 3 Operational amplifier U5B770931X (Fairchild)

OS1 Monostable multivibrator MC851P (Motorola)

OS2 Monostable multivibrator MC851P (Motorola)

OS3 Monostable multivibrator MC851P (Motorola)

OS4 Monostable multivibrator MC851P (Motorola)